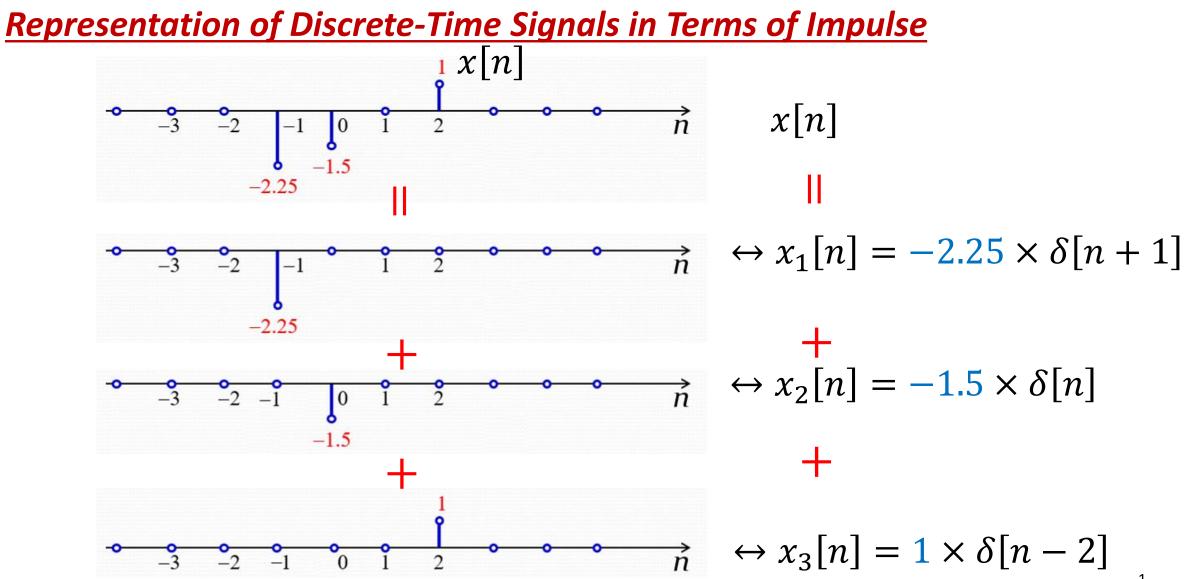
Linear Time-Invariant Systems (ch.2)

Discrete-Time LTI Systems

Continuous-Time LTI Systems

- **Properties of LTI Systems**
- **Differential or Difference Equations**

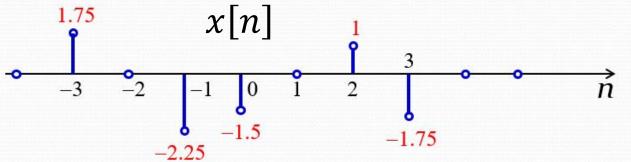






Representation of Discrete-Time Signals in Terms of Impulse

An arbitrary sequence can be represented as the weighted sum of shifted unit impulses



 $x[n] = 1.75\delta[n+3] - 2.25\delta[n+1] - 1.5\delta[n] + \delta[n-2] - 1.75\delta[n-3]$

□ A general form

$$x[n] = \sum_{k=-\infty}^{\infty} x[k]\delta[n-k]$$
 Sifting property of $\delta[n]$



Discrete-Time Unit Impulse Response and the Convolution-Sum

The response of a system to a unit impulse sequence $\delta[n]$ is called impulse response, denoted by h[n]





Discrete-Time Unit Impulse Response and the Convolution-Sum

- □ How to calculate the impulse response of a system
- For any system whose input-output relationship is defined by

$$y[n] = f\{x[n]\}$$

the impulse response h[n] is calculated as

 $h[n] = f\{\delta[n]\}$ replace x[n] by $\delta[n]$



Discrete-Time Unit Impulse Response and the Convolution-Sum

- □ How to calculate the impulse response of a system
- Examples: a system is defined as

$$y[n] = a_1 x[n] + a_2 x[n-1] + a_3 x[n-2] + a_4 x[n-3]$$

its impulse response h[n] is

$$h[n] = a_1 \delta[n] + a_2 \delta[n-1] + a_3 \delta[n-2] + a_4 \delta[n-3]$$



Discrete-Time Unit Impulse Response and the Convolution-Sum

- □ How to calculate the impulse response of a system
- Examples: a system is defined as n

$$y[n] = \sum_{k=-\infty}^{n} x[k]$$

its impulse response h[n] is

$$h[n] = \sum_{k=-\infty}^{n} \delta[k]$$



- □ How to calculate the impulse response of a system
- Examples: a system is defined as

$$y[n] = x_u[n-1] + \frac{1}{2}(x_u[n-2] + x_u[n])$$

its impulse response $h[n]$ is
$$h[n] = \delta[n-1] + \frac{1}{2}(\delta[n-2] + \delta[n])$$

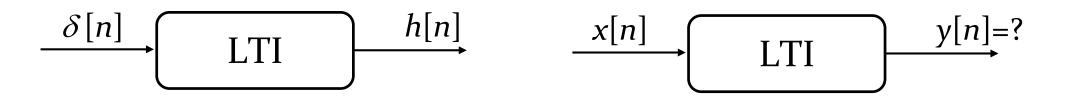


- An LTI discrete system is completely characterized by its impulse response
- In other words, knowing the impulse response one can compute the output of the LTI system for an arbitrary input



Discrete-Time Unit Impulse Response and the Convolution-Sum

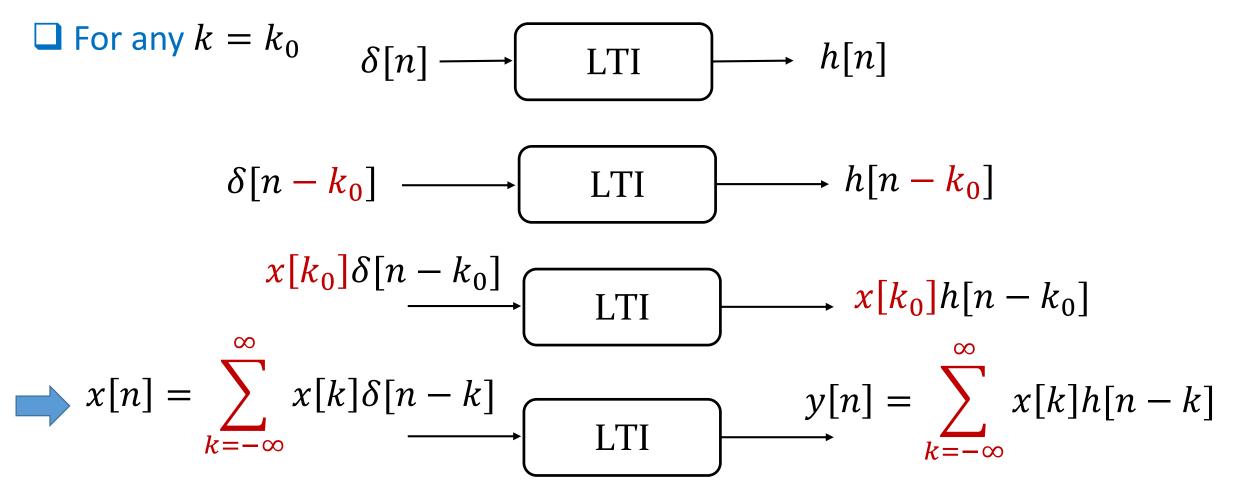
□ The impulse response completely characterizes an LTI system



Recall, an arbitrary input x[n] can be expressed as a linear combination of shifted unit impulses

$$x[n] = \sum_{k=-\infty}^{\infty} x[k]\delta[n-k]$$





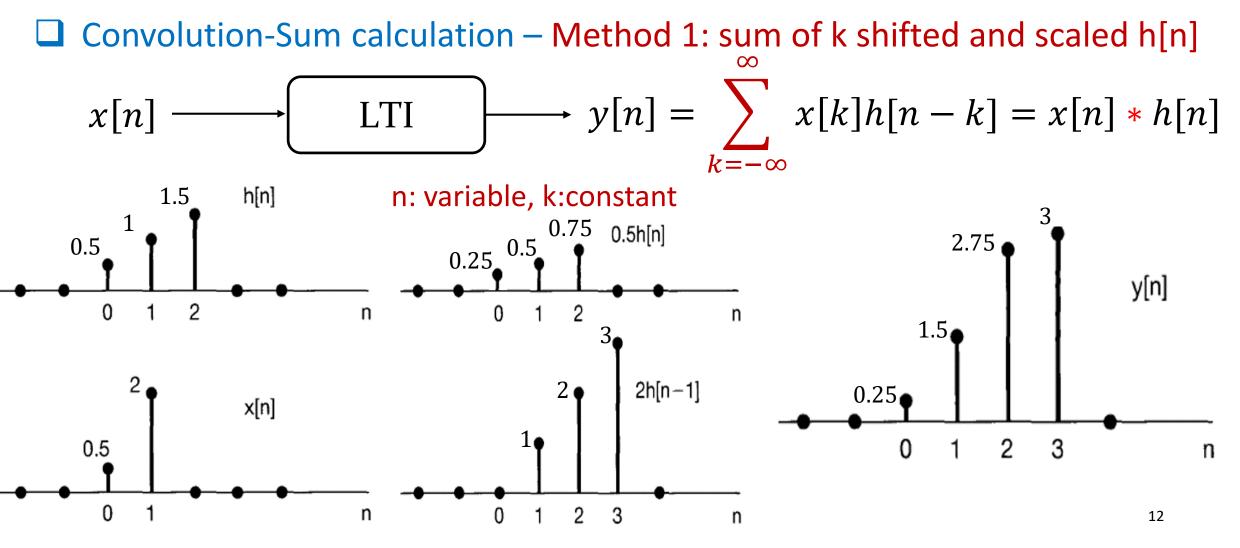


$$x[n] = \sum_{k=-\infty}^{\infty} x[k]\delta[n-k] \xrightarrow{\text{LTI}} y[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k]$$

$$\Box \sum_{k=-\infty}^{\infty} x[k]h[n-k] \text{ is refereed to as the convolution-sum}$$

$$x[n] \longrightarrow LTI \longrightarrow y[n] = x[n] * h[n]$$



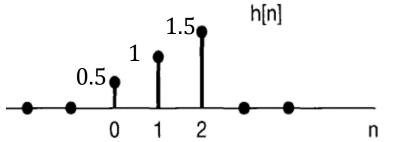


Discrete-Time Unit Impulse Response and the Convolution-Sum

Convolution-Sum calculation–Method 2: calculate y[n] for each n

$$y[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k] = x[n] * h[n]$$

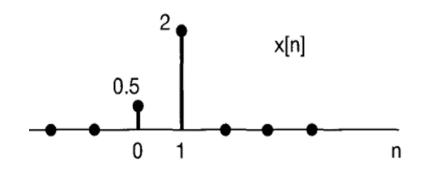
• Step 1: determine the range of k $k \in \{0,1\}$



• Step 2: determine the range of *n*

$$[n - k] \in \{0, 1, 2\} \leftrightarrow n \in \{0, 1, 2, 3\},$$

For other n, $y[n]=0$



Discrete-Time Unit Impulse Response and the Convolution-Sum

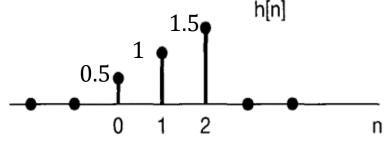
Convolution-Sum calculation–Method 2: calculate y[n] for each n $y[n] = \sum_{k=-\infty} x[k]h[n-k] = x[n] * h[n]$ ^{1.5}

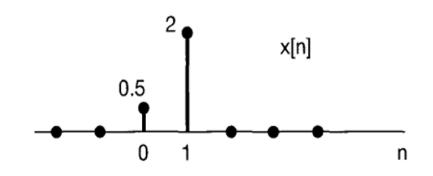
• Step 3: calculate y[n] for each n

$$y[\mathbf{0}] = \sum_{k=0}^{1} x[k]h[\mathbf{0} - k] = x[\mathbf{0}]h[\mathbf{0}] + x[\mathbf{1}]h[-1] = 0.25$$

$$y[1] = \sum_{k=0}^{1} x[k]h[1-k] = x[0]h[1] + x[1]h[0] = 1.5$$
$$y[2] = \sum_{k=0}^{1} x[k]h[2-k] = x[0]h[2] + x[1]h[1] = 2.75$$

$$y[\mathbf{3}] = \sum_{k=0}^{1} x[k]h[\mathbf{3} - k] = x[0]h[\mathbf{3}] + x[1]h[\mathbf{2}] = 3$$







Discrete-Time Unit Impulse Response and the Convolution-Sum

Convolution-Sum calculation–Method 3

$$y[n] = x[n] * h[n] = \sum_{k=-\infty} x[k]h[n-k]$$

 ∞

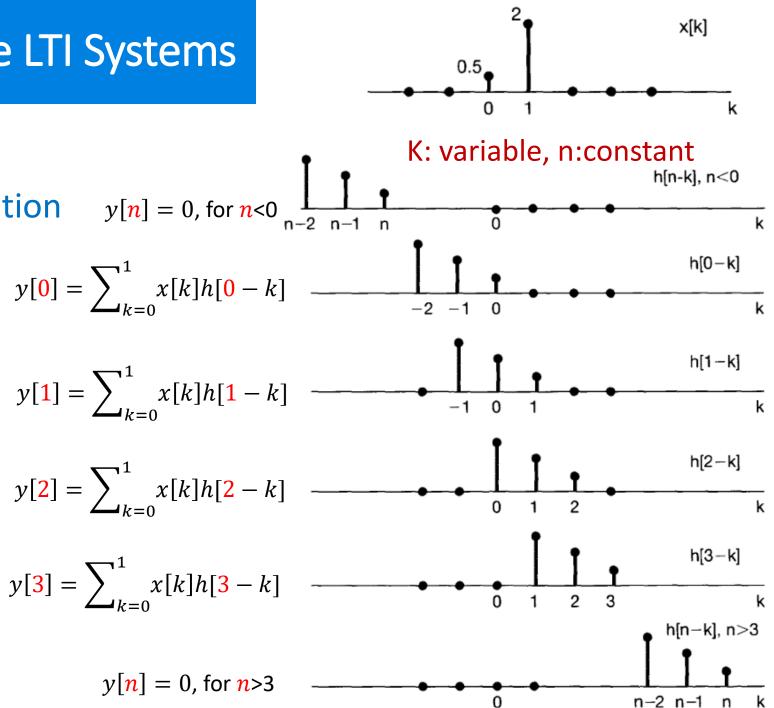
For each n :

- Step 1: change time variables $x[n] \rightarrow x[k], h[n] \rightarrow h[k],$ and reverse $h[k] \rightarrow h[-k]$
- Step 2: Shift $h[-k] \rightarrow h[n-k]$, n is considered as a constant
- Step 3: multiply $x[k] \cdot h[n-k]$
- Step 4: Summation $\sum_{k=-\infty}^{\infty} x[k] \cdot h[n-k]$

Change n, repeat step 1 to 4, calculate another y[n]

The Convolution-Sum

- Convolution-Sum calculation
 - Method 3
 - If the lengths of the two sequences are *M* and *N*, then the sequence generated by the convolution is of length *M*+*N*-1



The Convolution-Sum

Examples

$$y[\mathbf{n}] = x[n] * \delta[n] = \sum_{k=-\infty}^{\infty} x[k]\delta[\mathbf{n}-k] = x[n]$$

$$y[\mathbf{n}] = x[n] * \delta[n-d] = \sum_{k=-\infty}^{\infty} x[k]\delta[\mathbf{n}-k-d] \quad \text{Let } \mathbf{k} + \mathbf{d} = \mathbf{k'}$$

$$=\sum_{k'=-\infty}^{\infty}x[k'-d]\delta[n-k']$$

$$= x[n-d] * \delta[n] = x[n-d]$$

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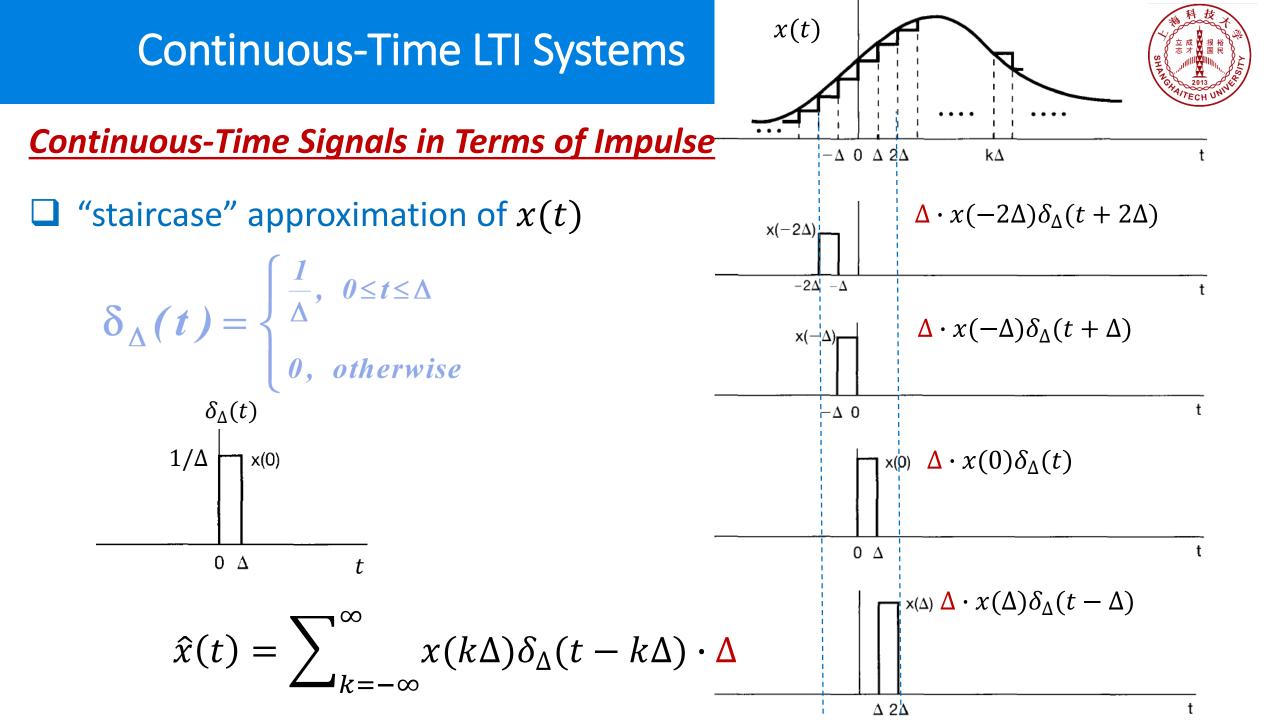


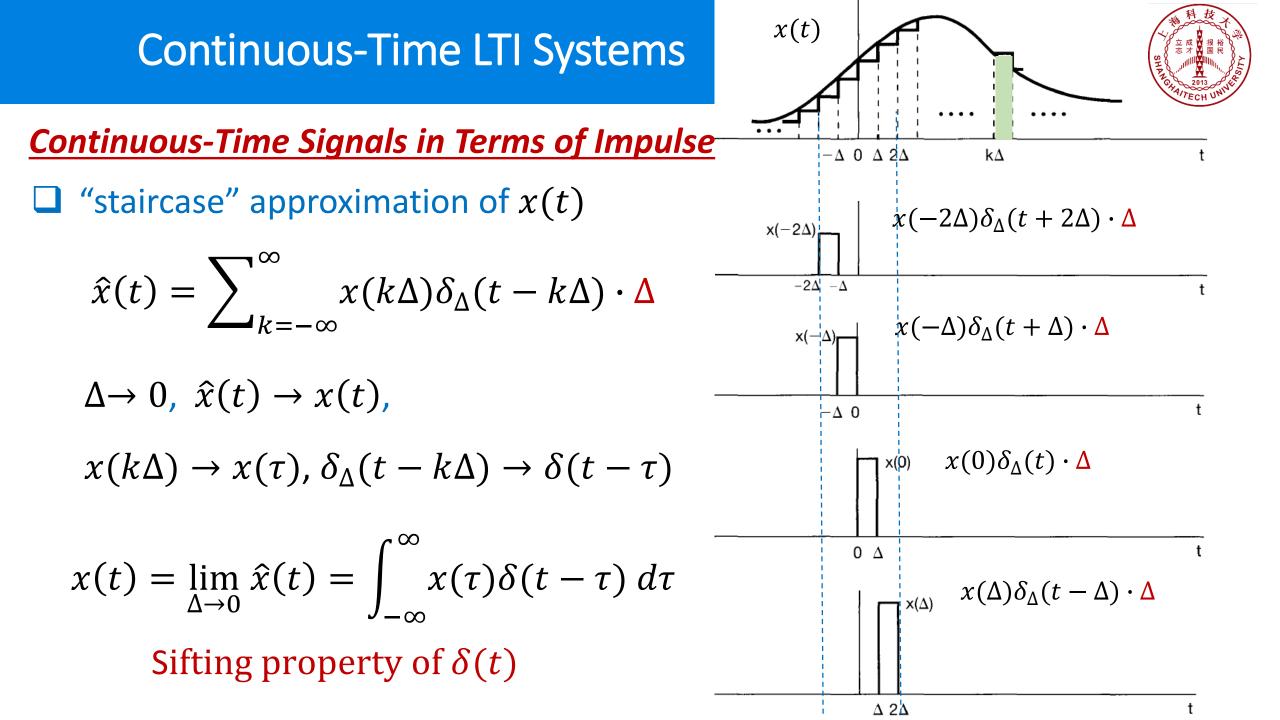
The Convolution-Sum Examples

$$y_{1}[n] = x[n] * h[n - m] = \sum_{k=-\infty}^{\infty} x[k]h[n - k - m] \quad \text{Let } k + m = k'$$
$$= \sum_{k'=-\infty}^{\infty} x[k' - m]h[n - k']$$
$$= x[n - m] * h[n] = y[n - m]$$

Linear Time-Invariant Systems (ch.2)

- Discrete-Time LTI Systems
- Continuous-Time LTI Systems
- Properties of LTI Systems
- Differential or Difference Equations





Continuous-Time Signals in Terms of Impulse

 \Box Using sampling property of $\delta(t)$

$$\int_{-\infty}^{\infty} x(\tau) \delta(t-\tau) \, d\tau = ?$$

$$x(t)\delta(t - t_0) = x(t_0)\delta(t - t_0)$$
 sampling property
$$x(\tau)\delta(t - \tau) = x(t)\delta(t - \tau)$$
 t: constant

$$\sum_{-\infty}^{\infty} x(\tau)\delta(t-\tau) d\tau = \int_{-\infty}^{\infty} x(t)\delta(t-\tau) d\tau \qquad x(\tau)\delta(t-\tau) \qquad x(t)$$

$$= x(t) \int_{-\infty}^{\infty} \delta(t-\tau) d\tau \qquad x(\tau)\delta(t-\tau) d\tau$$

$$= x(t) \qquad x(\tau)\delta(t-\tau) = x(\tau) \qquad x(\tau)\delta(t-\tau) \qquad x(\tau)$$

 $x(\tau)$

(a)

(b)

 $\delta(t-\tau)$

τ

Τ

t



Continuous-Time Signals in Terms of Impulse

An example

$$x(t) = \int_{-\infty}^{\infty} x(\tau) \delta(t-\tau) d\tau$$

$$u(t) = \int_{-\infty}^{\infty} u(\tau)\delta(t-\tau) d\tau = \int_{0}^{\infty} \delta(t-\tau) d\tau$$



Continuous-Time Unit Impulse Response and Convolution Integral

Continuous-Time Unit Impulse Response



What about

$$\begin{array}{c} x(t) & y(t) = ? \\ x(t) = \int_{-\infty}^{\infty} x(\tau)\delta(t-\tau) d\tau & y(t) = \int_{-\infty}^{\infty} x(\tau)h(t-\tau) d\tau
\end{array}$$

Sum of weighted and shifted impulses

Sum of weighted and shifted impulse response

$$\int_{-\infty}^{\infty} x(\tau)h(t-\tau) d\tau = x(t) * h(t)$$



Continuous-Time Unit Impulse Response and Convolution Integral

Computation convolution integral

$$x(t) * h(t) = \int_{-\infty}^{\infty} x(\tau)h(t-\tau) d\tau$$

- Change time variables $x(t) \to x(\tau)$, $h(t) \to h(\tau)$, and reverse $h(\tau) \to h(-\tau)$
- Shift $h(-\tau) \rightarrow h(t-\tau)$
- Multiply $x(\tau) \cdot h(t \tau)$
- Integral $\int_{-\infty}^{\infty} x(\tau) \cdot h(t-\tau) d\tau$

Continuous-Time Unit Impulse Response and Convolution Integral

Computation convolution integral: examples

$$x(t) * h(t) = \int_{-\infty}^{\infty} x(\tau)h(t-\tau) d\tau$$

$$x(t) * \delta(t) = \int_{-\infty}^{\infty} x(\tau) \delta(t-\tau) d\tau = x(t)$$

$$x(t) * \delta(t - t_0) = \int_{-\infty}^{\infty} x(\tau) \delta(t - \tau - t_0) d\tau = \int_{-\infty}^{\infty} x(\tau) \delta(t - (\tau + t_0)) d\tau$$

$$= \int_{-\infty}^{\infty} x(\tau' - t_0) \delta(t - \tau') d\tau' = x(t - t_0) * \delta(t)$$

$$= x(t-t_0)$$



Continuous-Time Unit Impulse Response and Convolution Integral

Computation convolution integral: examples

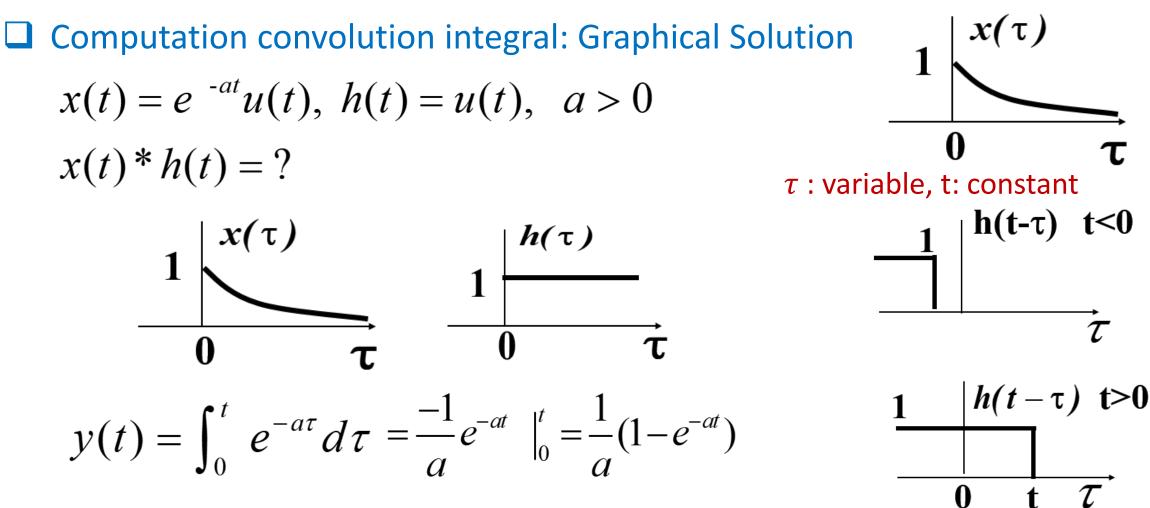
$$x(t) = e^{-at}u(t), h(t) = u(t), a > 0 x(t) * h(t) = ?$$

$$y(t) = \int_{-\infty}^{\infty} e^{-a\tau} u(\tau) \cdot u(t-\tau) d\tau$$

For
$$t < 0$$
 $x(\tau) \cdot h(t - \tau) = 0 \implies y(t) = 0$

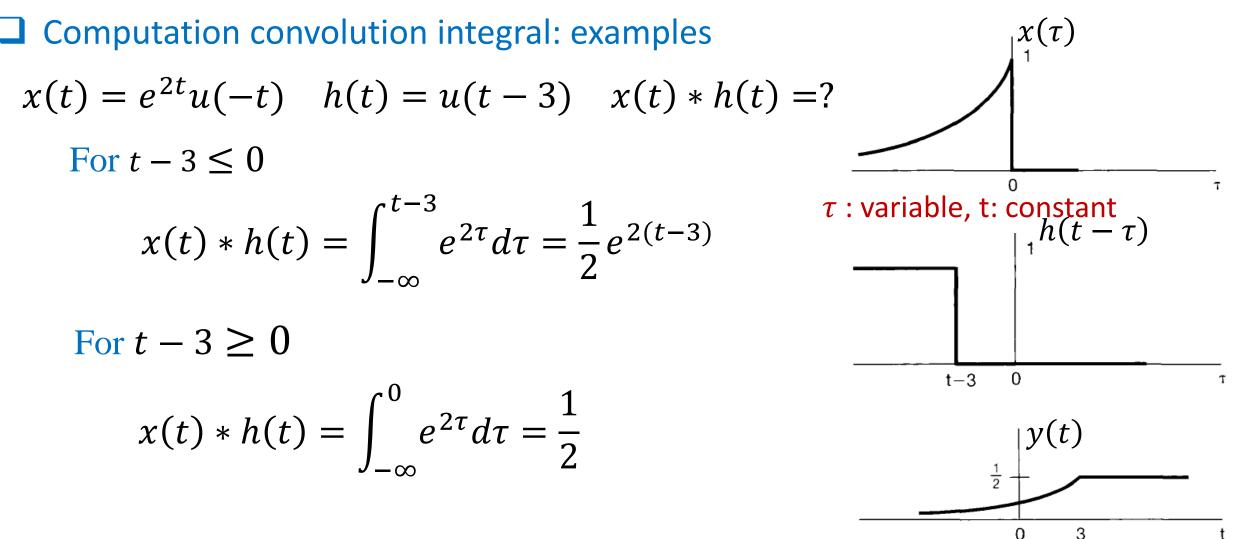
For
$$t \ge 0$$
 $y(t) = \int_0^t e^{-a\tau} d\tau = \frac{-1}{a} e^{-at} \Big|_0^t = \frac{1}{a} (1 - e^{-at})$

Continuous-Time Unit Impulse Response and Convolution Integral





Continuous-Time Unit Impulse Response and Convolution Integral

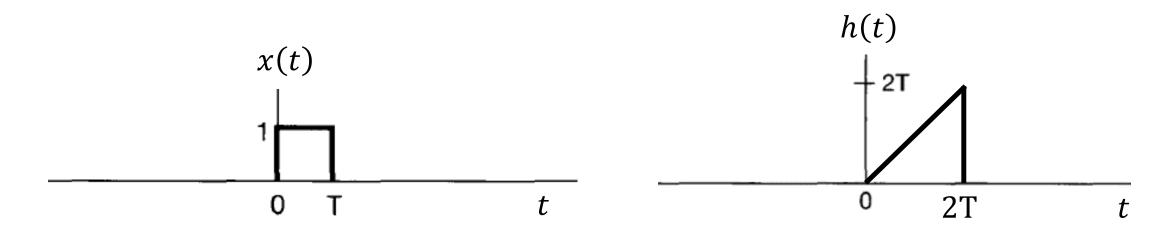




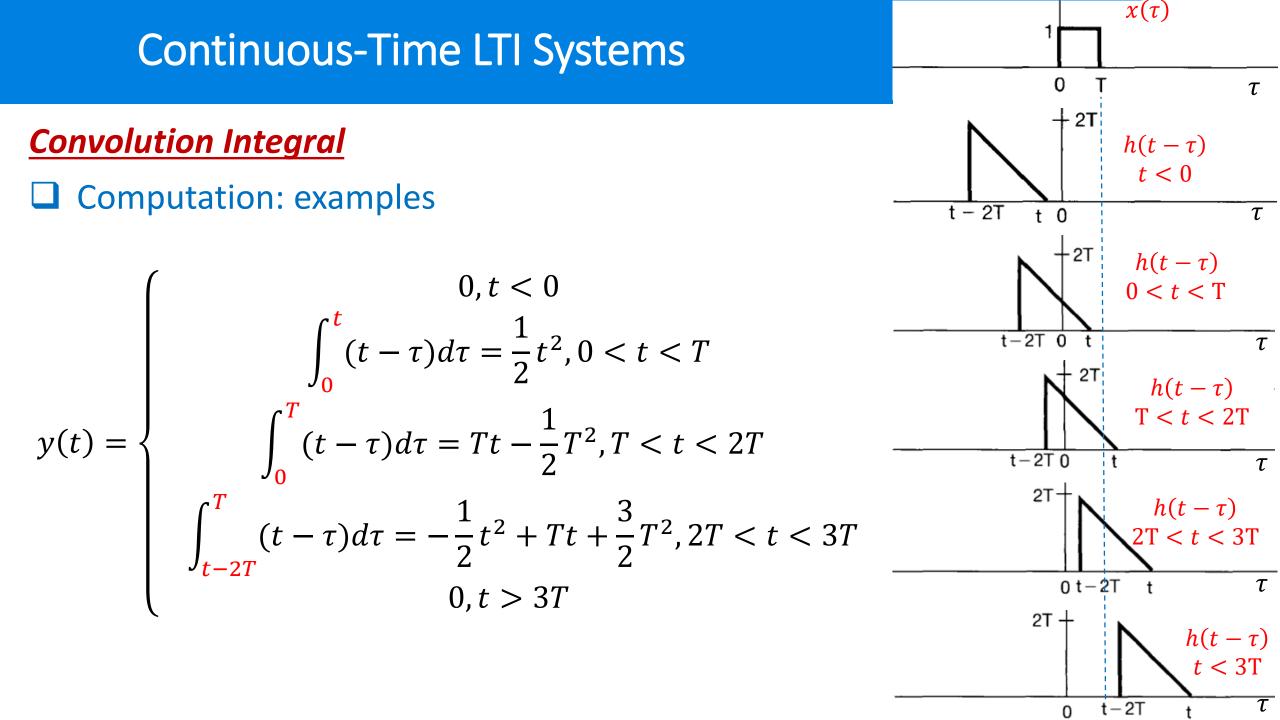
Continuous-Time Unit Impulse Response and Convolution Integral

Computation convolution integral: examples

$$x(t) = \begin{cases} 1, 0 < t < T \\ 0, \text{ otherwise} \end{cases} \quad h(t) = \begin{cases} t, 0 < t < 2T \\ 0, \text{ otherwise} \end{cases}$$



x(t) * h(t) = ?



Linear Time-Invariant Systems (ch.2)

- Discrete-Time LTI Systems
- Continuous-Time LTI Systems
- Properties of LTI Systems
- Differential or Difference Equations

Properties of LTI Systems

The commutative property

Discrete-time x[n] * h[n] = h[n] * x[n]

$$x[n] * h[n] = \sum_{k=-\infty}^{\infty} x[k]h[n-k] \stackrel{n-k}{=} m \sum_{m=-\infty}^{\infty} h[m]x[n-m] = h[n] * x[n]$$

Continuous-time x(t) * h(t) = h(t) * x(t)

$$x(t) * h(t) = \int_{-\infty}^{\infty} x(\tau)h(t-\tau) d\tau = \int_{-\infty}^{\infty} h(\tau')x(t-\tau')d\tau' = h(t) * x(t)$$

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Properties of LTI Systems

The distribute property

Discrete-time

$$x[n] * (h_1[n] + h_2[n]) = x[n] * h_1[n] + x[n] * h_2[n]$$
Proof

$$x[n] * (h_1[n] + h_2[n]) = \sum_{k=-\infty}^{\infty} x[k] (h_1[n-k] + h_2[n-k])$$

$$= \sum_{k=-\infty}^{\infty} x[k] h_1[n-k] + \sum_{k=-\infty}^{\infty} x[k] h_2[n-k])$$

 $= x[n] * h_1[n] + x[n] * h_2[n]$

Properties of LTI Systems

The distribute property

Continuous-time

$$x(t) * (h_1(t) + h_2(t)) = x(t) * h_1(t) + x(t) * h_2(t)$$

Proof

$$x(t) * (h_1(t) + h_2(t)) = \int_{-\infty}^{\infty} x(\tau) (h_1(t-\tau) + h_2(t-\tau)) d\tau$$

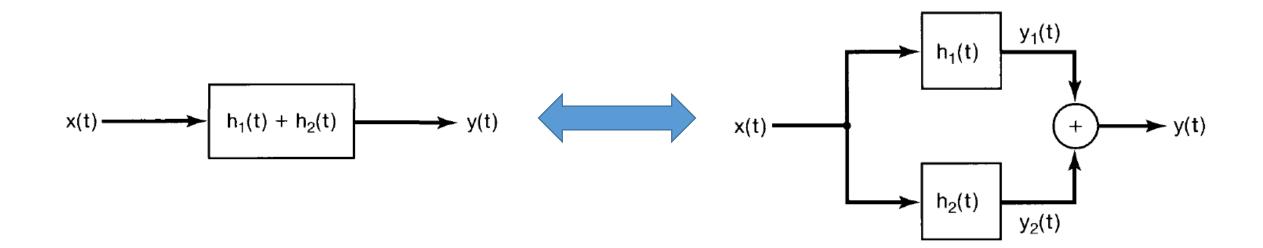
$$=\int_{-\infty}^{\infty} x(\tau)h_1(t-\tau)d\tau + \int_{-\infty}^{\infty} x(\tau)h_2(t-\tau)d\tau$$

$$= x(t) * h_1(t) + x(t) * h_2(t)$$

The distribute property

Continuous-time

$$x(t) * (h_1(t) + h_2(t)) = x(t) * h_1(t) + x(t) * h_2(t)$$



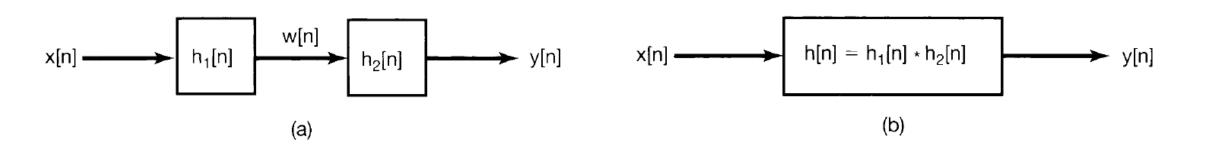


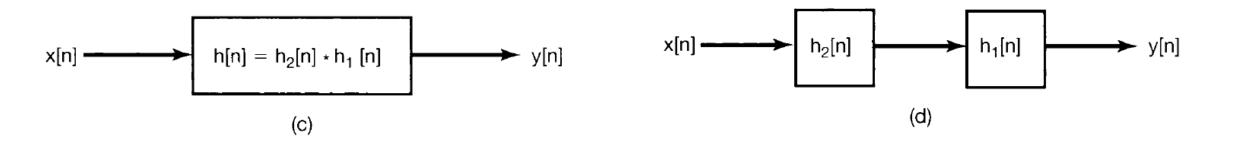
The associative property
Discrete-time
$$x[n] * (h_1[n] * h_2[n]) = (x[n] * h_1[n]) * h_2[n]$$

 $x[n] * (h_1[n] * h_2[n]) = x[n] * y[n], y[n] = \sum_{m=-\infty}^{\infty} h_1[m]h_2[n-m]$
 $= \sum_{k=-\infty}^{\infty} x[k]y[n-k] = \sum_{k=-\infty}^{\infty} x[k]\sum_{m=-\infty}^{\infty} h_1[m]h_2[n-k-m]$
Let $k + m = l$
 $= \sum_{k=-\infty}^{\infty} x[k]\sum_{l=-\infty}^{\infty} h_1[l-k]h_2[n-l]$
 $= \sum_{l=-\infty}^{\infty} \sum_{k=-\infty}^{\infty} x[k]h_1[l-k]h_2[n-l]$
 $= \sum_{l=-\infty}^{\infty} (x[l] * h_1[l])h_2[n-l] = (x[n] * h_1[n]) * h_2[n]$

The associative property

Discrete-time





The associative property

Continuous-time
$$x(t) * (h_1(t) * h_2(t)) = (x(t) * h_1(t)) * h_2(t)$$

 $x(t) * (h_1(t) * h_2(t)) = x(t) * \int_{-\infty}^{\infty} h_1(\tau)h_2(t - \tau)d\tau$
 $= \int_{-\infty}^{\infty} x(\tau') \int_{-\infty}^{\infty} h_1(\tau)h_2(t - \tau' - \tau)d\tau d\tau'$
Let $\tau' + \tau = \tau''$
 $= \int_{-\infty}^{\infty} x(\tau') \int_{-\infty}^{\infty} h_1(\tau'' - \tau')h_2(t - \tau'')d\tau'' d\tau'$
 $= \int_{-\infty}^{\infty} \int_{-\infty}^{\infty} x(\tau')h_1(\tau'' - \tau')d\tau' h_2(t - \tau'')d\tau''$
 $= \int_{-\infty}^{\infty} x(\tau'') * h_1(\tau'') h_2(t - \tau'')d\tau'' = (x(t) * h_1(t)) * h_2(t)$



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LTI systems with and without memory

□ Discrete-time system without memory only if h[n] = 0 for all $n \neq 0$ $h[n] = h[0]\delta[n] = k\delta[n]$ y[n] = kx[n] Why?

- **Continuous-time system without memory only if** h(t) = 0 for all $t \neq 0$
 - $h(t) = h(0)\delta(t) = k\delta(t) \qquad y(t) = kx(t)$

Invertibility for LTI systems

□ If $h_0(t) * h_1(t) = \delta(t)$, the system $h_1(t)$ is the inverse of the system $h_0(t)$

$$x(t)$$
 $h_0(t)$ $h_1(t)$ $w(t)=x(t)$

 \Box Similarly, if $h_0[n]\ast h_1[n]=\delta[n]$, the system $h_1[n]$ is the inverse system of $h_0[n]$

Invertibility for LTI systems

Examples

Consider $h_0[n] = u[n]$, determine the inverse system $h_1[n]$

$$\therefore h_0[n] * h_1[n] = u[n] * h_1[n] = \delta[n]$$

$$\delta[n] = u[n] - u[n-1] = u[n] * (\delta[n] - \delta[n-1])$$

$$\therefore h_1[n] = \delta[n] - \delta[n-1]$$

Invertibility for LTI systems

Examples

Consider the LTI system consisting of a pure time shift

$$y(t) = x(t - t_0),$$

determine the inverse system.

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Causality for LTI systems

□ If h[n] = 0 for n < 0, or h(t) = 0 for t < 0, the system is causal

Equivalent to the condition of initial rest

$$y[n] = \sum_{k=-\infty}^{n} x[k]h[n-k] \quad \text{or} \quad y[n] = \sum_{k=0}^{\infty} h[k]x[n-k]$$

$$\int_{0}^{t} \int_{0}^{\infty} \int_{0}^{\infty} f_{n}^{0}$$

$$y(t) = \int_{-\infty} x(\tau)h(t-\tau) d\tau \quad \text{or} \quad y(t) = \int_{0}^{\infty} h(\tau)x(t-\tau)d\tau$$

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Properties of LTI Systems

Causality for LTI systems

Examples

• Accumulator: $y[n] = \sum_{l=-\infty}^{n} x[l]$ Causal LTI system

$$h[n] = \sum_{l=-\infty}^{n} \delta[l] = u[n] \qquad h[n] = 0 \text{ for } n < 0$$

• Factor 2 interpolator: $y[n] = x_u[n] + \frac{1}{2}(x_u[n-1] + x_u[n+1])$

Non-Causal LTI system

$$h[n] = \delta[n] + \frac{1}{2}(\delta[n-1] + \delta[n+1])$$
$$h[n] \neq 0 \text{ for } n = -1$$



Stability for LTI systems

A discrete LTI system is stable if *h*[*n*] is absolutely summable

 \Box A continuous LTI system is stable if h(t) is absolutely integrable

$$\sum_{k=-\infty}^{\infty} |h[k]| < \infty \qquad \text{absolutely summable}$$

 $\int_{-\infty}^{\infty} |h(\tau)| d\tau < \infty \qquad \text{absolutely integrable}$



Stability for LTI systems

□ Proof: "if and only if" (Sufficient and necessary condition)

$$|y[n]| = \left| \sum_{k=-\infty}^{\infty} h[k]x[n-k] \right| \le \sum_{k=-\infty}^{\infty} |h[k]x[n-k]| = \sum_{k=-\infty}^{\infty} |h[k]| |x[n-k]|$$

$$\therefore |y[n]| \le \sum_{k=-\infty}^{\infty} |h[k]| |x[n-k]|$$

$$|f|x[n-k]| \le B_x \qquad |y[n]| \le B_x \sum_{k=-\infty}^{\infty} |h[k]|$$

If and only if $\sum_{k=-\infty}^{\infty} |h[k]| < \infty$ $|y[n]| < \infty$

 $|\mathbf{f}|_{\mathcal{H}}(\mathbf{f}) = \mathbf{h}(\mathbf{f}) = \mathbf{h}(\mathbf{f})$

Stability for LTI systems

Proof: continuous case

$$|y(t)| = \left| \int_{-\infty}^{\infty} h(\tau) x(t-\tau) d\tau \right| \le \int_{-\infty}^{\infty} |h(\tau)| \cdot |x(t-\tau)| d\tau \le B_x \int_{-\infty}^{\infty} |h(\tau)| d\tau$$

If and only if $\int_{-\infty}^{\infty} |h(\tau)| d\tau < \infty$ $|y(t)| < \infty$

Stability for LTI systems

Examples

$$y[n] = x[n - n_0]$$

$$h[n] = \delta[n - n_0]$$

$$\sum_{n=-\infty}^{\infty} |h[n]| = \sum_{n=-\infty}^{\infty} |\delta[n-n_0]| = 1$$



Stability for LTI systems

Examples $h[n] = \alpha^n \mu[n]$

$$\sum_{n=-\infty}^{\infty} |h[n]| = \sum_{n=-\infty}^{\infty} |\alpha^n| \mu[n]| = \sum_{n=0}^{\infty} |\alpha^n| = \frac{1}{1-|\alpha|} \quad \text{If } |\alpha| < 1$$

If $|\alpha| = 1$, the system is unstable

The unit step response of LTI systems

The unit step response, s(t) or s[n], corresponding to the output with input x(t) = u(t) or x[n] = u[n]

$$s[n] = \mu[n] * h[n] = \sum_{k=-\infty}^{\infty} h[k]u[n-k] = \sum_{k=-\infty}^{n} h[k]$$

$$\mu[n] = \sum_{k=-\infty}^{n} \delta[k] \qquad \qquad s[n] = \sum_{k=-\infty}^{n} h[k]$$

$$h[n] = s[n] - s[n-1]$$



The unit step response of LTI systems

The unit step response, s(t) or s[n], corresponding to the output with input x(t) = u(t) or x[n] = u[n]

$$s(t) = \mu(t) * h(t) = \int_{-\infty}^{\infty} h(\tau)u(t-\tau)d\tau = \int_{-\infty}^{t} h(\tau)d\tau$$

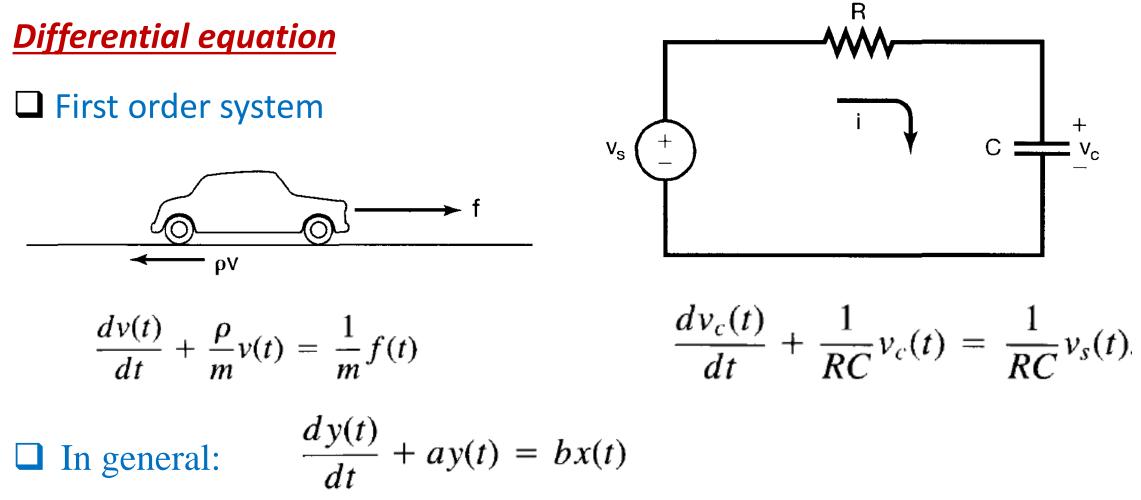
$$\mu(t) = \int_{-\infty}^{t} \delta(\tau) d\tau \qquad s(t) = \int_{-\infty}^{t} h(\tau) d\tau$$

$$h(t) = \frac{ds(t)}{dt} = s'(t)$$

Linear Time-Invariant Systems (ch.2)

- Discrete-Time LTI Systems
- Continuous-Time LTI Systems
- **Properties of LTI Systems**
- Differential or Difference Equations





Describes a relationship between the input and the output (implicit)
Auxiliary conditions are needed for solving the DE.

Differential equation

□ First order system: example

$$\frac{dy(t)}{dt} + 2y(t) = x(t)$$

$$If x(t) = Ke^{3t}u(t) \qquad y(t) = ?$$

Solution:

$$y(t) = y_p(t) + y_h(t)$$

 $y_p(t)$: particular solution, *forced response (same form as input)*

 $y_h(t)$: Homogenous solution

$$\frac{dy(t)}{dt} + 2y(t) = 0$$



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Differential equation

□ First order system: example

$$\frac{dy(t)}{dt} + 2y(t) = x(t)$$

If $x(t) = Ke^{3t}u(t)$ y(t) = ?Particular solution: Let $y_p(t) = Ye^{3t}$, for t > 0

$$3Ye^{3t} + 2Ye^{3t} = Ke^{3t} \implies Y = K/5 \implies y_p(t) = \frac{K}{5}e^{3t}$$

U Homogenous solution: Let $y_h(t) = Ae^{st}$, for t>0

$$Ase^{st} + 2Ae^{st} = 0 \implies s = -2 \implies y_h(t) = Ae^{-2t}$$

$$y(t) = Ae^{-2t} + \frac{K}{5}e^{3t}$$
, for $t > 0$

Differential equation

$$y(t) = Ae^{-2t} + \frac{K}{5}e^{3t}$$
, for $t > 0$

Auxiliary condition is required to determine A

□ Initial rest as auxiliary condition for causal LTI systems: y(0) = 0

$$A + \frac{K}{5} = 0 \implies A = -\frac{K}{5} \implies y(t) = \frac{K}{5} (e^{3t} + e^{-2t}), \text{ for } t > 0$$
$$= \frac{K}{5} (e^{3t} + e^{-2t})u(t)$$



Differential equation

General case: Nth-order linear constant-coefficient differential equation

$$\sum_{k=0}^{N} a_k \frac{d^k y(t)}{dt^k} = \sum_{k=0}^{M} b_k \frac{d^k x(t)}{dt^k}$$

D Particular solution + Homogenous solution: $y(t) = y_p(t) + y_h(t)$

- $y_p(t)$: forced response (same form as input)
- $y_h(t)$: Natural response, $\sum_{k=0}^N a_k \frac{d^k y(t)}{dt^k} = 0$

Initial rest as auxiliary condition, that is if x(t) = 0 for $t \le t_0$,

$$y(t_0) = \frac{dy(t_0)}{dt} = \dots = \frac{d^{N-1}y(t_0)}{dt^{N-1}} = 0$$

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Difference equation

General case: Nth-order linear constant-coefficient difference equation

$$\sum_{k=0}^{N} a_{k} y[n-k] = \sum_{k=0}^{M} b_{k} x[n-k]$$

D Particular solution + Homogenous solution: $y[n] = y_p[n] + y_h[n]$

- $y_p[n]$: forced response (same form as input)
- $y_h[n]$: Natural response, $\sum_{k=0}^N a_k y[n-k] = 0$

□ Initial rest as auxiliary condition, that is if x[n] = 0 for $n \le n_0$,

$$y[n_0] = y[n_0-1] = \dots = y[n_0-(N-1)] = 0$$

Difference equation

□ Recursive solution:

$$y[n] = \frac{1}{a_0} \left\{ \sum_{k=0}^{M} b_k x[n-k] - \sum_{k=1}^{N} a_k y[n-k] \right\}$$

• Particular case *N*=0

$$y[n] = \frac{1}{a_0} \sum_{k=0}^{M} b_k x[n-k]$$
 Non-recursive equation

$$h[n] = \frac{1}{a_0} \sum_{k=0}^{M} b_k \delta[n-k]$$

Finite impulse response (FIR) system

Difference equation

□ Recursive solution: example

$$y[n] - \frac{1}{2}y[n-1] = x[n]$$

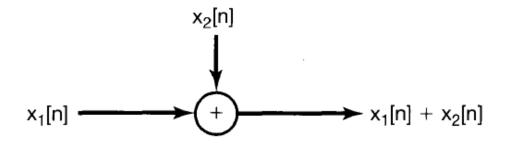
• Consider $x[n] = K\delta[n]$ and take initial rest: y[-1] = 0

$$y[0] = x[0] + \frac{1}{2}y[-1] = K \qquad y[1] = x[1] + \frac{1}{2}y[0] = \frac{1}{2}K$$
$$y[2] = x[2] + \frac{1}{2}y[1] = \left(\frac{1}{2}\right)^2 K \qquad \dots \qquad y[n] = x[n] + \frac{1}{2}y[n-1] = \left(\frac{1}{2}\right)^n K$$
$$\therefore h[n] = \left(\frac{1}{2}\right)^n u[n] \qquad \text{Infinite impulse response (IIR) system}$$

Generally $\sum_{k=0}^{N} a_k y[n-k] = \sum_{k=0}^{M} b_k x[n-k] \begin{cases} N = 0, \text{ FIR system} \\ N > 0, \text{ IIR system} \end{cases}$ Not always!

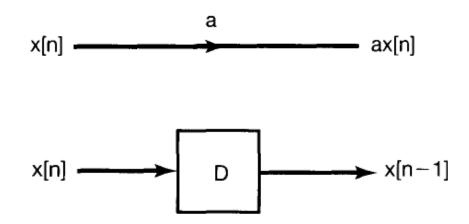
Block Diagram Representations

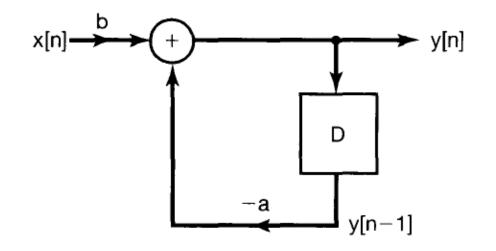
□ Basic elements: discrete-time



$$y[n] + ay[n-1] = bx[n]$$

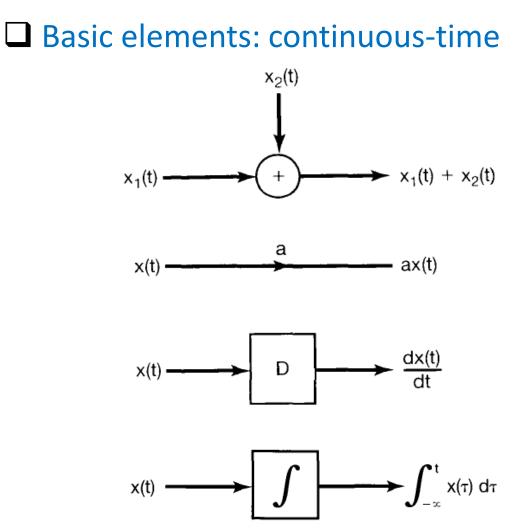
$$y[n] = -ay[n-1] + bx[n]$$





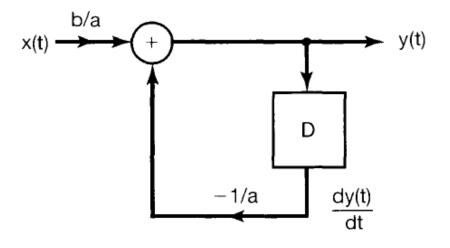


Block Diagram Representations



$$\frac{dy(t)}{dt} + ay(t) = bx(t)$$

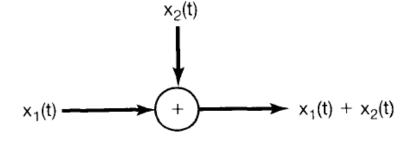
$$y(t) = -\frac{1}{a}\frac{dy(t)}{dt} + \frac{b}{a}x(t)$$

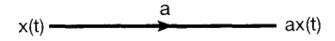


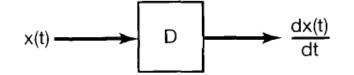


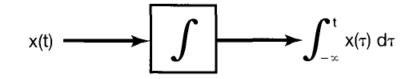
Block Diagram Representations











$$\frac{dy(t)}{dt} + ay(t) = bx(t)$$
$$\frac{dy(t)}{dt} = -ay(t) + bx(t)$$
$$y(t) = \int_{-\infty}^{t} [bx(\tau) - ay(\tau)]d\tau$$

